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Abstract: Lubricants are of key importance for mechanical processing, which exists in nearly every 11 mechanical system. When the equipment is in operation, debris particles will be generated in me-12 chanical lubricants. The detection of debris particles can reflect the wear degree of machinery com-13 ponents, and further provide risk pre-warning of the system before the fault occurs. In this work, a 14 novel type of inductive debris sensor consisting of two excitation coils and two sensing coils is pro-15 posed. The developed sensor is proved to be of high sensitivity through the verification of experi-16 ments. The testing results show that, using the designed sensor, the ferrous metal debris with the 17 size of 115 μ m and nonferrous metal debris with the size of 313 μ m in the pipe with an inner diam-18 eter of 12.7 mm can be well detected, respectively. Moreover, the proposed inductive debris sensor 19 structure has better sensitivity at higher throughput and its design provides a useful insight into the 20 development of high-quality sensors with superior performances. 21

Keywords: inductive debris sensor, online debris monitoring, metal debris

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Copyright: © 2021 by the authors. Submitted for possible open access publication under the terms and conditions of the Creative Commons Attribution (CC BY) license (https://creativecommons.org/license s/by/4.0/). 1. Introduction

Fault detection and condition monitoring of machines are quite important methods 25 to maintain the operational performance and extend the service life of rotating and recip-26 rocating machinery in many sectors such as machinery manufacturing, transportation in-27 dustry, and military. The applications of these technologies can prevent the breakdown 28 of critical system components and avoid unexpected production delays [1]. Detecting 29 metal debris in the lubrication oil is a direct and dependable method for monitoring the 30 condition of rotating and reciprocating machinery [2-4]. Under normal operating condi-31 tions, the metal debris retains a stable size and concentration in the lubrication oil. How-32 ever, when there is abnormality then the concentration and size of metal debris will in-33 crease [5-7]. Taking into account this situation, the working condition real-time monitor-34 ing of mechanical equipment has attracted increasing attention from researchers. Since 35 the real-time online detection of metal debris in the lubricating oil is a important task, 36 several new techniques and methods have been developed over the past recent decades 37 to improve the accuracy of debris detection. 38

In general, the existing detection techniques including various online and off-line 39 inspection methods, can be divided into the following six classes: optical scattering counter method [8, 9], capacitance method [10], resistance method [11], ultrasonic method [12], 41 X-ray method [13], and inductive method [14-16]. Different detection methods have different advantages, and undoubtedly have some limitations which constrain their industrial utilizations. For example, the reliability of the optical method is kind of poor because 45 it requires the transparency of both the oil and inclusive bubbles. The application of 45

capacitance or resistance methods will induce oil deterioration, which will degrade the 46 detection accuracy as time goes. The accuracy of the ultrasonic method is affected by the 47 viscosity of the oil, the flow rate, and mechanical vibration which is hard to be eliminated 48 in practical applications. The X-ray method is with high detection precision, but the de-49 mand on complex equipment is indispensable. For the inductive method, it is suitable for 50 both metal and non-metal pipelines and the associated equipment is in a simple structure. 51 Moreover, the sensitivity of this method does not rely on the oil quality, and it can effec-52 tively distinguish non-ferrous and ferrous metal debris. However, it has some certain lim-53 itations including the low sensitivity to non-ferrous metal debris and the incapability of 54 detecting debris shape. From the practical point of view, the inductive method is the most 55 feasible and effective technique for many applications. 56

Since the inductive method has many advantages, a lot of studies have been con-57 ducted by researchers in this field. Flanagan et al. [17] first proposed a method for testing 58 debris material and size with a single-coil sensor in 1990. Experimental results showed 59 that the sensor can effectively detect debris of 100 µm in a pipe with a 6-mm diameter. In 60 industrial applications, MetalSCAN from GasTop is a widely used sensor. It consists of 61 one induction coil and two excitation coils around the same tube. The specifications of the 62 MetalSCAN product indicate that its sensitivity to ferrous and non-ferrous metal debris 63 in the inner diameter of the pipe, which was approximate 9.525 mm [18], could be 64 achieved with values of 100 μ m and 405 μ m, respectively. One problem that remains to 65 be solved is that the detection performance of this sensor is seriously affected by the back-66 ground noise and vibration signals. Talebi etal.[19] designed the sensor to effectively de-67 tect 125 µm ferrous debris in pipes with an internal diameter of 4mm, and it can detect 68 the concentration of metal debris in the oil. However, the 4mm-diameter of the pipe limits 69 the flow rate of the oil. In order to improve the accuracy of detection, Ren et al. [20] pro-70 posed a sensor using an excitation coil and two induction coils. It can identify the 120 µm 71 ferrous debris and 210 µm non-ferrous debris in a 34 mm-diameter pipe. However, the 72 induction coil should be immersed into the oil, which will result in the increased resistance 73 to the flow of lubricants. Du et al. [21-23] made improvements on the original basis of the 74 sensor using the parallel LC resonance method. The sensor's sensitivity was obviously 75 improved with the ability to detect the 20 µm debris. Its excellent performance benefited 76 from the use of a microfluidic channel with a diameter of 250 µm. The practical application 77 of this sensor is still limited because the micro-size of the channel leads to the blockage. 78 Also, a considerable throttling effect which results in the unsuitability of the sensor to 79 high-rate flow tests, exists in the channel. 80

In order to develop a high-sensitive sensor that is suitable for the high-rate flow test, a novel sensor design consisting of two excitation coils and two sensing coils has been proposed in this paper. To prove the sensitivity and applicability of the developed sensor, experimental tests have been conducted to demonstrate its superior performance.

2. Sensor principle design



Figure 1. The structure of the new designed inductive debris sensor.

The mechanical structure of the sensor is mainly composed of two excitation coils 88 and two sensing coils. The two sensing coils are placed side by side, with two sides being 89 symmetrical, and the two excitation coils are arranged right outside the two sensing coils 90 respectively, as shown in Figure 1. 91

The sensor's operating principle is as shown in Figure 2. An AC voltage is applied to 92 the excitation coils, which generates the magnetic field as shown in Figure 2(a). When 93 ferrous metal debris enters the sensor, two factors (permeability and eddy current) will 94 interact with each other, as shown in Figure 2(b). First, the magnetic flux will increase due 95 to the higher permeability of the ferrous metal debris. Second, a magnetic field whose 96 direction is opposite to the original magnetic field will be generated by the eddy currents 97 inside the ferrous metal debris, which will decrease the total magnetic flux. At low fre-98 quency, the increase of magnetic flux dominates, which means a positive voltage pulse 99 will be generated when ferrous metal debris flows through the sensor. 100



Figure 2. The magnetic field distribution of the sensor-designed sensor: (**a**) no metal debris flows through; (**b**) when ferrous metal debris enters the sensor.

3. Mathematical modeling of sensors

According to Biot-Savart's theorem, the magnetic field of a circular current-carrying 106 wire is[24, 25] 107

$$B = \frac{\mu_0 I r^2}{2 \left(r^2 + x^2\right)^{\frac{3}{2}}} \tag{1}$$

Where B is the magnetic field strength of the circular current-carrying wire at the108target point, μ_0 is the vacuum magnetic permeability, I is the excitation current, r is109the radius of the circle, and x is the transverse coordinate of the target point.110

The sensor's parameter model is shown in Figure 3. Where n_1 is the number of turns 111 per unit length of the excitation coil, R_1 is the inner diameter of the excitation coil, R_2 112 is the outer diameter of the excitation coil, N_1 is the number of turns of the excitation 113 coil, R is the inner diameter of the sensing coil, N_2 is the number of turns of the sensing coil, I is the amplitude of the excitation signal, L is the length of the sensing coil, 115 the midpoint of the excitation coil is set as the origin, x is the axial distance. 116

Generally, inductive sensors are composed of multiple layers of solenoids. The central axis of the solenoid is set as the origin. The magnetic field at any point on the axis of the multi-layer solenoid is represented as follows. 120

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Figure 3. The diagram of the sensor.

$$B = \frac{\mu_0 n_1 I}{2} \left[-\left(x - \frac{L}{2}\right) \ln \frac{R_2 + \sqrt{R_2^2 + \left(x - \frac{L}{2}\right)^2}}{R_1 + \sqrt{R_1^2 + \left(x - \frac{L}{2}\right)^2}} + \left(x + \frac{L}{2}\right) \ln \frac{R_2 + \sqrt{R_2^2 + \left(x + \frac{L}{2}\right)^2}}{R_1 + \sqrt{R_1^2 + \left(x + \frac{L}{2}\right)^2}} \right] (2)$$

Assuming that the metal debris are spherical with radius r_a , the change in axial magnetic flux when metal debris enters the sensor[23] 124

$$d\varphi = dB \cdot S = (\mu_r - 1)\pi R^2 B V_0 \tag{3}$$

Where V_0 is the volume of the metal debris, $V_0 = 4/3\pi r_a^3$, according to the 125 princple of electromagnetic induction can be obtained from the sensing coil generated by 126 the induction electromotive force is 127

$$E = -N_2 \left(\mu_r - 1\right) \pi R^2 V_0 \frac{dB}{dt} \tag{4}$$

The excitation signal is a sinusoidal AC current $i = I \cos(2\pi ft)$, and the induced 128 electric potential is 129

$$E = -N_2 \left(\mu_r - 1\right) \pi R_1^2 V_0 \frac{\mu_0 n_1}{2} \left(i \frac{dK}{dt} + K \frac{di}{dt}\right)$$
(5)

Where

$$K = \left[-\left(x - \frac{L}{2}\right) \ln \frac{R_2 + \sqrt{R_2^2 + \left(x - \frac{L}{2}\right)^2}}{R_1 + \sqrt{R_1^2 + \left(x - \frac{L}{2}\right)^2}} + \left(x + \frac{L}{2}\right) \ln \frac{R_2 + \sqrt{R_2^2 + \left(x + \frac{L}{2}\right)^2}}{R_1 + \sqrt{R_1^2 + \left(x + \frac{L}{2}\right)^2}} \right]$$
(6)

Assuming that the velocity of the metal debris through the sensor is v, the position 131 of the metal debris is x = vt - L/2, then the induced electric potential is 132

$$E_{1} = -\frac{2N_{1}N_{2}(\mu_{r}-1)\mu_{0}\pi^{2}R_{1}^{2}r_{a}^{3}I}{3L(R_{2}-R_{1})}\left[\cos(2\pi ft)\frac{dK}{dt} - 2\pi fK\sin(2\pi ft)\right]$$
(7)

Since the two sets of coils have the same structure, when the metal debris passes 133 through the second set of coils, the induced electric potential is 134

$$E_{2} = -\frac{2N_{1}N_{2}(\mu_{r}-1)\mu_{0}\pi^{2}R_{1}^{2}r_{a}^{3}I}{3L(R_{2}-R_{1})}\left[\cos\left(2\pi f(t-\Delta t)\right)\frac{dK}{dt} - 2\pi fK\sin\left(2\pi f(t-\Delta t)\right)\right]$$
(8)

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The time difference between the metal debris passing through the two sets of coils is $\Delta t = (L+d)/v$, where *d* is the distance between the two sets of coils. The induced 136 electric potential output from the sensor is 137

$$E = E_1 - E_2 \tag{9}$$

We can obtain the curve of the induced electrostatic force according to Equation 9 as shown in Figure 4.



Figure 4. The curve of induction electromotive force of mathematical model.

4. Experimental process

4.1. Design of the sensor

The manufacturing process of the designed four-coil structure is briefly introduced 140 below. First, make the sensing coil and wind 0.1 mm diameter enameled wire on an epoxy 141 resin skeleton with an inner diameter of 12.7 mm and a thickness of 1 mm (because the 142 magnetic permeability of epoxy resin is close to that of air, the epoxy resin has a small 143 effect on the magnetic field), with a total of 4-layer winding and 200 turns per layer. Then, 144 make the excitation coil by winding 0.2 mm diameter enameled wire around the outside 145 of the sensing coil, with a total of 4-layer and 100 turns per layer. 146

4.2. Signal processing method

In order to extract the accurate response signal of metal debris, and reduce the high-148 frequency noise disturbance to a minimum degree, the output voltage of the sensor sens-149 ing coil, a simple and effective signal acquisition, and a processing system are designed 150 in our work, as shown in Figure 5. A sinusoidal signal of ±10 V and 125 kHz is generated 151 as the excitation signal of the sensor system (Through experiments, we know that the sen-152 sor has the highest sensitivity when the excitation frequency is 120 – 130 kHz, this will be 153 confirmed later). In the sensing coil, a sinusoidal signal with the same frequency as the 154 excitation signal is then induced. When metal debris passes through the sensing area, a 155 signal will be generated correspondingly, which, however, is very weak and emerges with 156 the induced sinusoidal signal. The variation of the signal arising from the metal debris is 157 hard to be detected directly so that a signal processing system is needed. 158

Firstly, the AC signal is converted to a DC signal by true RMS conversion(The "true 159 RMS conversion" means the process in which the full-wave rectification of a sinusoidal 160 signal is followed by low-pass filtering, then the signal is converted to a DC signal). The 161 DC signal is then differentially amplified using the low-noise amplifier INA114 with a 162 gain of G=500(The INA114 is made by Texas Instruments, Inc). Due to errors arisen in coil 163 processing, the two DC signals are slightly different. After differential amplification, there 164 presents a nonzero signal called bias voltage, which will affect the next step of the ampli-165 fication effect. A compensation voltage (Ve) is introduced during the second amplification 166

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to balance the bias voltage. Still, INA114 low-noise amplifier is adopted, with gain G=100.167It uses a Chebyshev filter to improve the signal-to-noise ratio. The oscilloscope shows the168final output results. A signal similar to a cycle of a sine function will be detected when169metal debris passes through the sensing area of the sensor.170



Figure 5. Sensor signal acquisition and processing system.

4.3. Experimental setup

The schematic diagram of the experimental platform is shown in Figure 6. To be able 174 to accurately control the speed and position of the metal debris passing through the sensor 175 area. The metal debris can be fixed in the nylon rope. Besides, the nylon rope is driven by 176 a motor, and the moving speed of the nylon rope is controlled by controlling the speed of 177 the motor, then controlling the speed and position of the metal debris through the sensor 178area. The nylon's permeability is close to that of air, so the nylon rope has a small effect 179 on the magnetic field. In the practical case, the shapes of metal debris produced by the 180 mechanical wear process are not consistent, which causes difficulty for experimental anal-181 ysis. In order to better quantify the experimental results, in our work, nearly spherical 182 metal debris is used in the experiment. 183



Figure 6. Schematic diagram of the experimental platform.

5. Experimental results and discussion

5.1. Experimental result

For experimental comparison study, a series of ferrous metal debris is selected, with diameters of 150 μ m, 200 μ m, 250 μ m and 300 μ m respectively (the tolerance is approximately ±10 μ m). The excitation signal is ±10 V and 125 kHz. The velocity of the metal

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debris passing through the sensor is fixed as 0.2 m/s. The final output signals of the corre-191 sponding metal debris are shown in Figure 7. The first graph shows the noise level of the 192 sensor without metal debris passing through. An obvious output signal (greater than the 193 background noise voltage) can be observed when ferrous metal debris with a diameter of 194 150 µm passes through the sensor, which means the designed sensor can well detect the 195 ferrous metal debris with a diameter larger than 150 µm. The amplitude of output voltage 196 correspondingly increases with the increase of the diameter of metal debris. The relation-197 ship between metal debris size and the output voltage is shown in Figure 8 (where each 198 metal debris size is counted using 12 sets of experimental data, and the short line indicates 199 the standard deviation), and the output voltage is proportional to the volume of the metal 200 debris as can be derived from Equation (7). Based on this law, we can determine the size 201 of the metal debris by detecting the output voltage value. Since the output voltage signal 202 is proportional to the debris volume, it can be deduced that the detection limit of the sen-203 sor is 150 μ m ÷ $\frac{3}{880}/400 \approx 115 \mu$ m(The magnitude of the noise included in the circuit is 204 400 mV, and the magnitude of the output voltage is 880 mV when a ferrous metal debris 205 with a diameter of 150 µm passes through the sensor). 206



Figure 7. Voltage signals are generated by the passage of ferrous metal debris of different diame-
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Figure 8. Variation of output voltage with metal debris size.

5.2. Sensor's frequency characteristic

For inductive sensors, the excitation frequency is also one of the key factors affecting 213 the sensitivity of the sensor. A group of experiments is carried out to study the influence 214 of excitation frequency on the sensor's sensitivity, which elected 300um ferrous metal de-215 bris for the experiment. The speed of metal debris passing through the sensor is still fixed 216 as 0.2 m/s, and the excitation signal voltage is $\pm 10 \text{ V}$. The experimental results are shown 217 in Figure 9 (All the experiments are repeated 12 times, and the values shown in the figure 218 take an average of the 12 tested values, and the short line indicates the standard deviation). 219 The experimental results show the sensor's sensitivity reaches the maximum when the 220 excitation frequency of 120 - 130 kHz. 221



Figure 9. The sensor's frequency characteristic.

5.3. Influence of radial distribution of the magnetic field on sensitivity

Since the magnetic field inside the tube excited by the excitation coils is non-uniform 225 in the radial direction, the output voltages will be different when the passing through 226 metal debris present at different radial positions, which will lead to inaccurate estimation 227 of the metal debris. The magnetic field distribution of the sensor is simulated by COMSOL 228 software, and the result is shown in Figure 10.(In Figure 10, the two sets of excitation coils 229 are wound in opposite directions. The plane perpendicular to the axis of the coil is taken 230 as the Z=0 plane at the midpoint of a set of excitation coils.) We can easily verify the non-231 uniform distribution of the magnetic field in the radial direction. Bo is the magnetic flux 232

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density at z=0 and r=0 (with the center of the specific excitation coils as origin). B(r) repre-233 sents the magnetic flux density along the r direction in the plane of z=0. In Figure 11, the 234 relationship between relative magnetic flux density B(r)/B₀ and the location on r direction 235 is given. It can be inferred that the maximum measurement error of the sensor is about 236 10%. For experimental verification, a 300 µm ferrous metal debris is selected, with the 237 same velocity but at different radial positions. The test results are shown in Figure 12. Vo 238 is the voltage output when metal debris passes through the center of the sensor. It can be 239 seen that the error caused by the difference in the radial position is within 12% (Due to 240 the existence of error in the experimental process, resulting in a certain difference between 241 the experimental results and simulation results). 242



Figure 10. The magnetic flux density distribution of an excitation coil.



Figure 11. Radial distribution of relative magnetic flux density at z=0.

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Figure 12. The output voltage relative to r=0 value when metal debris pass through different radial positions.

5.4. Influence of the axial distribution of metal debris on the output voltage.

During the operation of machinery and equipment, more than one metal debris is 251 produced. When the spacing between two metal debris is too short, the voltages they gen-252 erate will be superimposed, making it difficult to recognize the true size of the metal de-253 bris. Two metal debris of the same size were selected for the experiment and passed 254 through the sensor with different spacing and the same speed (0.2 m/s), and the output 255 results are shown in Figure 13. The induced voltages of adjacent debris at different inter-256 vals are shown in Figure 13. From the experimental results, it is obvious that when the 257 spacing is less than 25 mm, the output voltage signals are completely superimposed to-258 gether, and when the spacing is greater than 90 mm, the output voltage signals are com-259 pletely separated. 260



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Figure 13. Induced voltage for two metal debris at different distances: (**a**) 15 mm; (**b**) 25 mm; (**c**) 40 267 mm; (**d**) 55 mm; (**e**) 70 mm; (**f**) 90 mm. 268

5.5 Sensor's speed characteristic.

To verify the effect of the speed of metal debris passage on the sensitivity of the sensor. We select 200 μ m ferrous metal debris for the experiment. Similarly, the excitation 271 signal is ± 10 V and 125 kHz. The metal debris passes through the sensor at different speeds, 272 and the experimental output is shown in Figure 14(the short line indicates the standard 273 deviation). We can see from the experimental results that the faster the metal debris passes 274 through the sensor, the greater the voltage amplitude of the sensor output and the higher 275 the sensitivity of the sensor. 276



Figure 14. Voltage signals are generated by the passage of nonferrous metal debris of different diameters.

5.6 Nonferrous debris detection sensitivity

The ability of the sensor to detect nonferrous magnetic metal debris was also verified. 281 Copper debris with diameters of 500 µm and 800 µm were selected for the experiments. 282 Similarly, the excitation signal is ±10 V and 125 kHz. The velocity of the copper debris 283 passing through the sensor is fixed as 0.2 m/s. The final output signals of the correspond-284 ing copper debris are shown in Figure 15. The experimental results clearly indicate the 285 output signal is in opposite phase to the ferrous particle signal. Therefore, the type of par-286 ticle can be identified by observing the signal phase. Assuming that the output signal am-287 plitude is proportional to the volume of the debris, it can be deduced that the detection 288 limit of the sensor for nonferrous is 500 μ m ÷ $\sqrt[3]{1360/400} \approx 313 \mu$ m(The noise level of the 289 circuit is 400 mV, and the output voltage is 1360 mV when a copper debris with a diameter 290 of 500 µm passes through the sensor). 291

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Figure 15. Voltage signals are generated by the passage of nonferrous metal debris of different diameters.

6. Conclusions

In this paper, a novel sensor structure with dual-excitation and dual-sensing coils has 296 been proposed for online debris monitoring. With the successful fabrication of such prin-297 ciple prototype, ferrous metal debris with a diameter of 115 µm and nonferrous metal 298 debris with a diameter of 313 µm can be detected using the sensor probe with the diameter 299 of 12.7 mm. To enable the senor with better detection capability of metal debris, effects of 300 the excitation frequency and radial distribution of the magnetic field on sensor sensitivity 301 have been investigated. Results show that the highest sensitivity of the sensor has been 302 achieved with the excitation frequency in the range of 120 to 130 kHz. Also, the radial non-303 uniform distribution of the magnetic field has remarkably influenced the detection accu-304 racy by up to 12%. Furthermore, distance distribution of metal debris along the axial di-305 rection on the voltage output has been discussed. It is worth noting that the output voltage 306 signal is completely separable when the distance between two particles is greater than 307 70mm. In summary, the proposed sensor design has the ability to produce a more stable 308 waveform and the superior performance of such device has been demonstrated through-309 out the experimental tests in term of the high sensitivity. This novel sensor design also 310 provides a useful insight into the development of high-quality sensors with superior per-311 formances. In future research, design optimization of the sensor will be conducted to im-312 prove the detection stability and precision. 313

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